

LM831 Low Voltage Audio Power Amplifier

General Description

The LM831 is a dual audio power amplifier optimized for very low voltage operation. The LM831 has two independent amplifiers, giving stereo or higher power bridge (BTL) operation from two- or three-cell power supplies.

The LM831 uses a patented compensation technique to reduce high-frequency radiation for optimum performance in AM radio applications. This compensation also results in lower distortion and less wide-band noise.

The input is direct-coupled to the LM831, eliminating the usual coupling capacitor. Voltage gain is adjustable with a single resistor.

Features

- Low voltage operation, 1.8V to 6.0V
- High power, 440 mW, 8Ω, BTL, 3V
- Low AM radiation
- Low noise
- re- Low THD

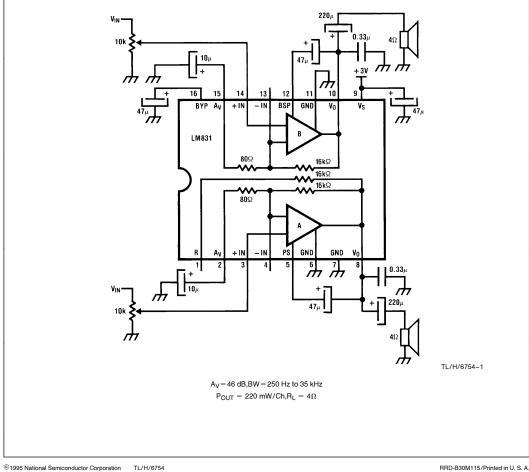
Applications

- Portable tape recorders
- Portable radios
- Headphone stereo
- Portable speakers

Typical Application

Dual Amplifier with Minimum Parts

LM831 Low Voltage Audio Power Amplifier



December 1994

Absolute Maximum Rat	tings		
If Military/Aerospace specified de please contact the National Se Office/Distributors for availability a	miconductor Sales	Storage Temperature, T _{stg} Junction Temperature, T _j Lead Temp. (Soldering, 10 sec.), T _l	−65°C to +150°C +150°C +260°C
Supply Voltage, V _S	7.5V	Thermal Resistance	
Input Voltage, V _{IN}	\pm 0.4V	$\theta_{\rm JC}$ (DIP)	27°C/W
Power Dissipation (Note 1), PD	1.3W (M Package)	θ_{JA} (DIP)	75°C/W
	1.4W (N Package)	$\theta_{\rm JC}$ (SO Package)	20°C/W
Operating Temperature (Note 1), $\mathrm{T}_{\mathrm{opr}}$	-40°C to +85°C	θ_{JA} (SO Package)	95°C/W

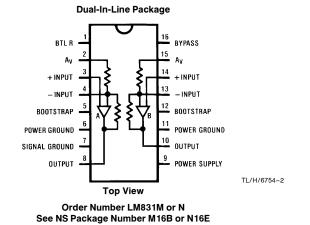
Electrical Characteristics

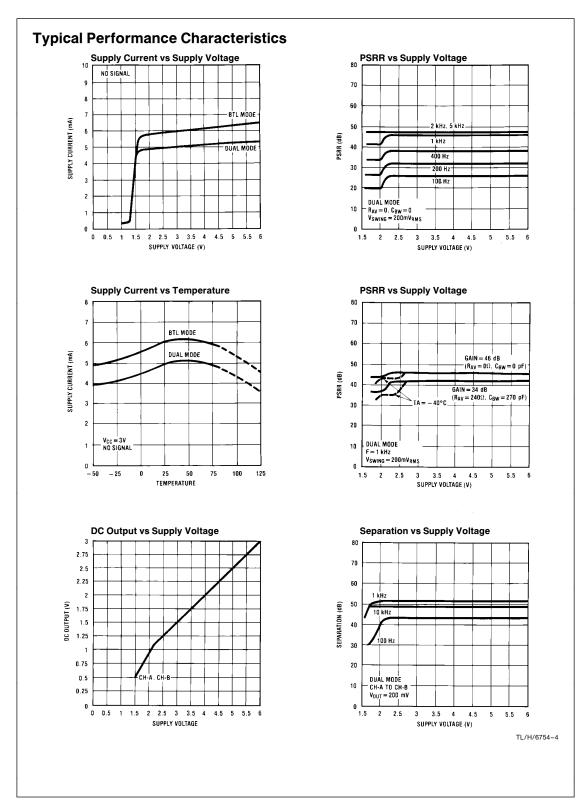
Unless otherwise specified, T_A = 25°C, V_S = 3V, f = 1 kHz, test circuit is dual or BTL amplifier with minimum parts.

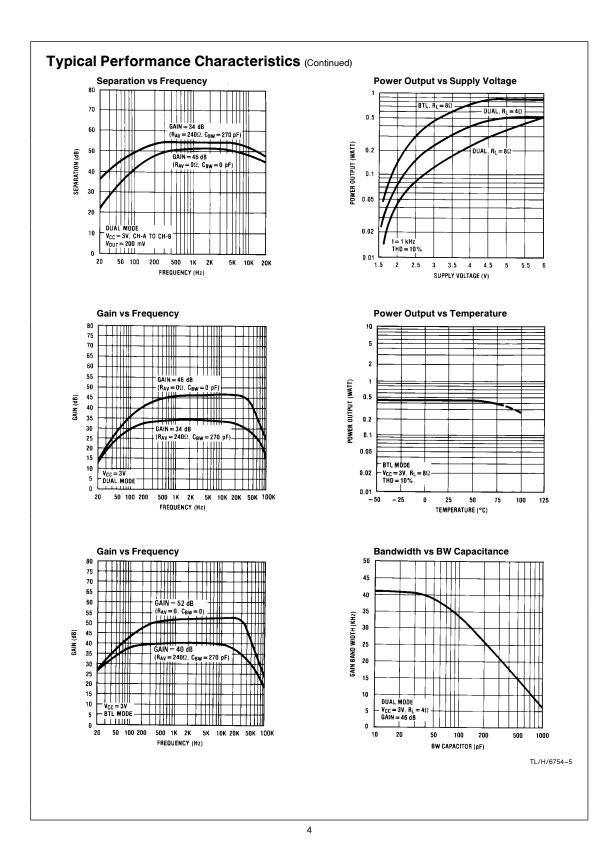
Symbol	Parameter	Conditions	Тур	Tested Limit	Unit (Limit)
V _S	Operating Voltage		3 3	1.8 6	V(Min) V(Max)
IQ	Supply Current	$V_{IN} = 0$, Dual Mode $V_{IN} = 0$, BTL Mode	5 6	10 15	mA (Max) mA (Max)
V _{OS}	Output DC Offset	V _{IN} = 0, BTL Mode	10	50	mV (Max)
R _{IN}	Input Resistance		25	15 35	k (Min) k (Max)
A _V	Voltage Gain	$V_{IN} = 2.25 \text{ mV}_{rms}$, f = 1 kHz, Dual Mode	46	44 48	dB (Min) dB (Max)
PSRR	Supply Rejection	$V_{S} = 3V + 200 \text{ mV}_{rms} @ f = 1 \text{ kHz}$	46	30	dB (Min)
P _{OD}	Power Out	$V_S = 3V, R_L = 4\Omega,$ 10% THD, Dual Mode	220	150	mW (Min)
P _{ODL}	Power Out Low, V _S	$V_{S} = 1.8V, R_{L} = 4\Omega,$ 10% THD, Dual Mode	45	10	mW (Min)
P _{OB}	Power Out	$V_{S} = 3V, R_{L} = 8\Omega,$ 10% THD, BTL Mode	440	300	mW (Min)
P _{OBL}	Power Out Low, V _S	$V_{S} = 1.8V, R_{L} = 8\Omega,$ 10% THD, BTL Mode	90	20	mW (Min)
Sep	Channel Separation	Referenced to $V_0 = 200 \text{ mV}_{rms}$	52	40	dB (Min)
IB	Input Bias Current		1	2	μΑ (Max)
E _{n0}	Output Noise	Wide Band (250 \sim 35 kHz)	250	500	μV (Max)
THD	Distortion	$V_{S} = 3V, P_{O} = 50 \text{ mW},$ f = 1 kHz, Dual	0.25	1	% (Max)

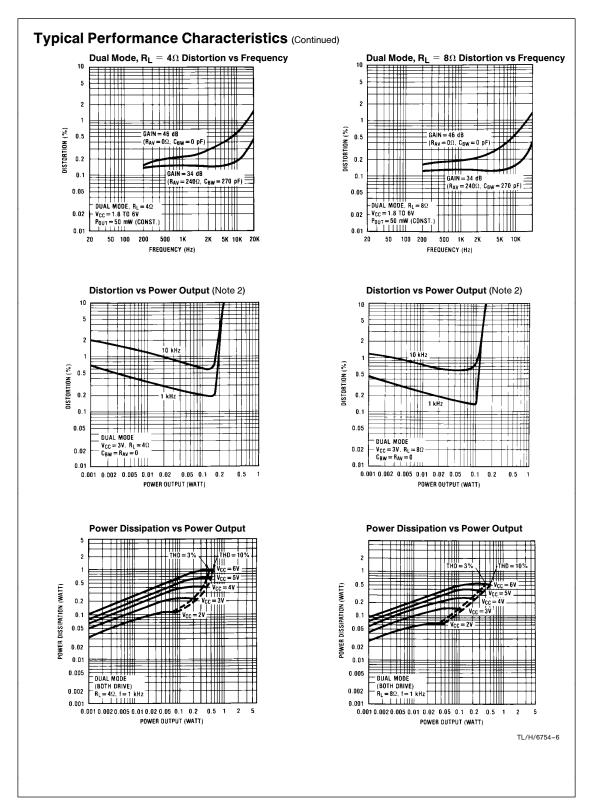
Note 1: For operation in ambient temperatures above 25°C, the device must be derated based on a 150°C maximum junction temperature and a thermal resistance of 98°C/W junction to ambient for the M package or 90°C/W junction to ambient for the N package.

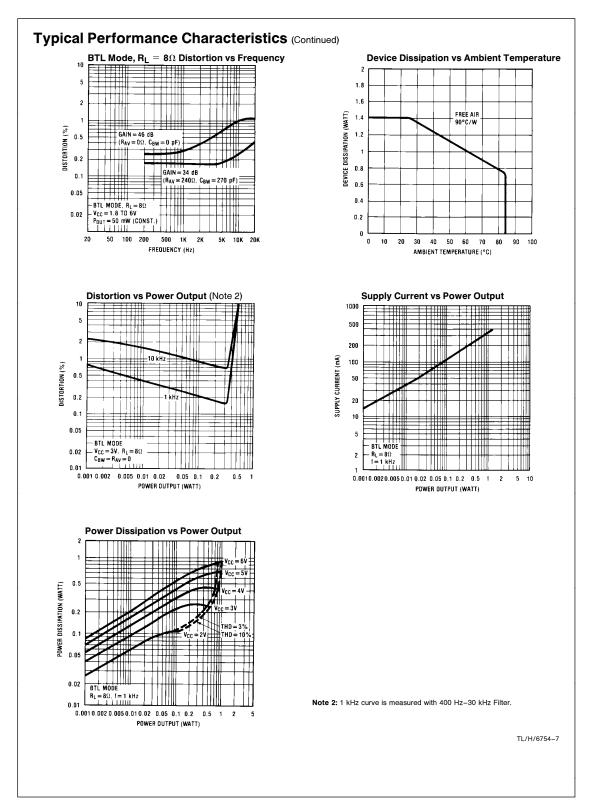
Connection Diagram

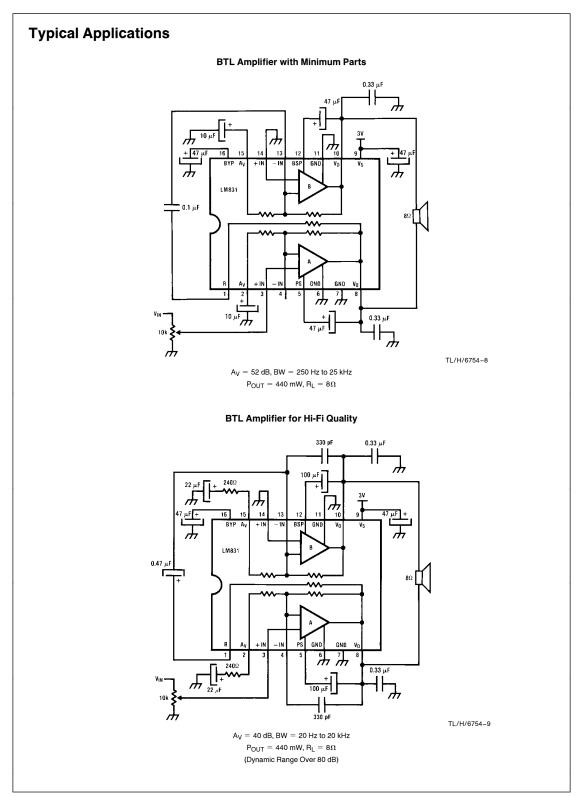


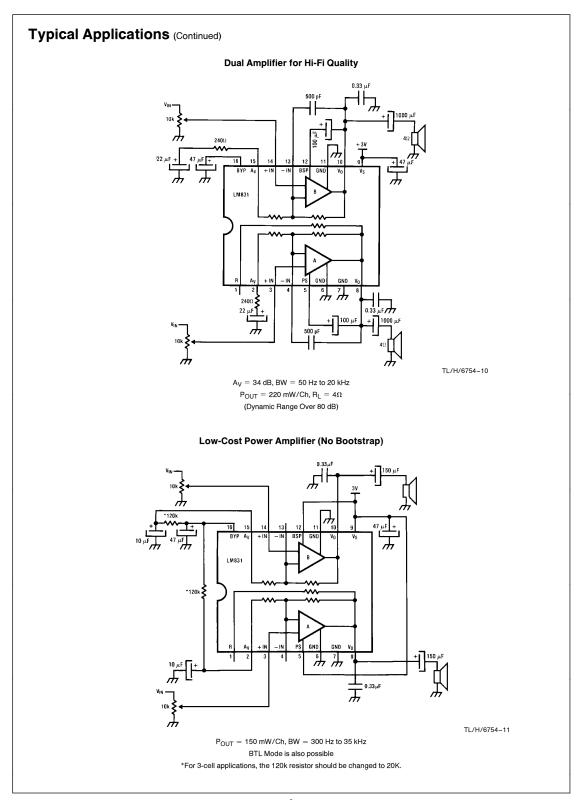












LM831 Circuit Description Refer to the external component diagram and equivalent schematic.

The power supply is applied to Pin 9 and is filtered by resistor R₁ and capacitor C_{BY} on Pin 16. This filtered voltage at Pin 16 is used to bias all of the LM831 circuits except the power output stage. Resistor R₀ generates a biasing current that sets the output DC voltage for optimum output power for any given supply voltage.

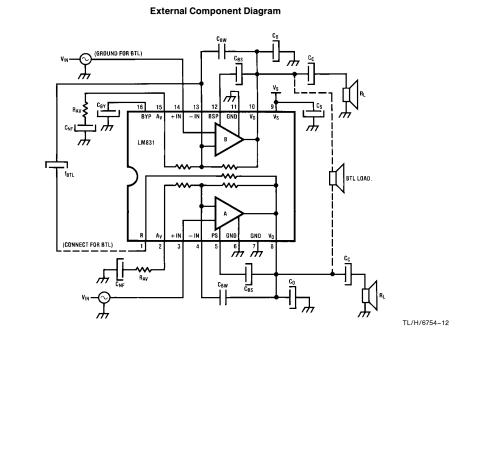
The capacitor $C_{\mbox{NF}}$ on Pin 2 provides unity DC gain for maximum DC accuracy.

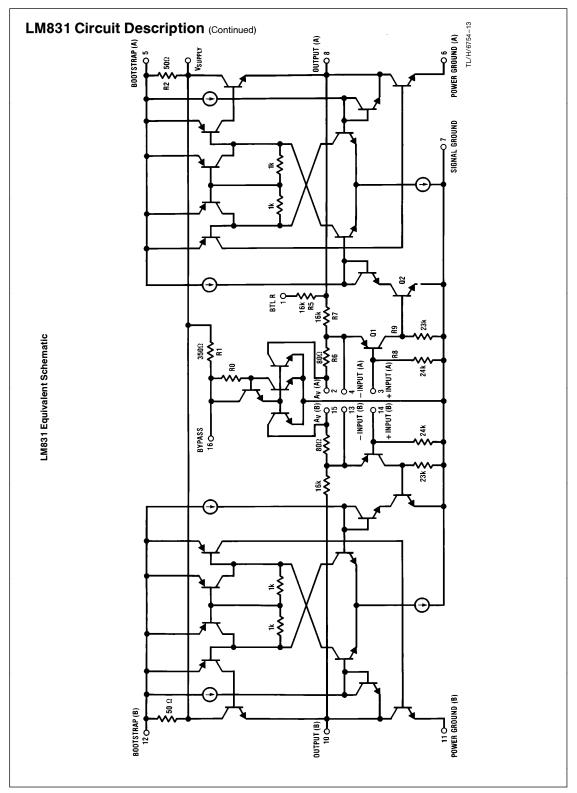
 Q_2 provides voltage gain and the rest of the devices buffer the output load from $Q_2{\,}{}^{\prime}{}^{\prime}{}^{\prime}$ collector.

Bootstrapping of Pin 5 by C_{BS} allows maximum output swing and improved supply rejection.

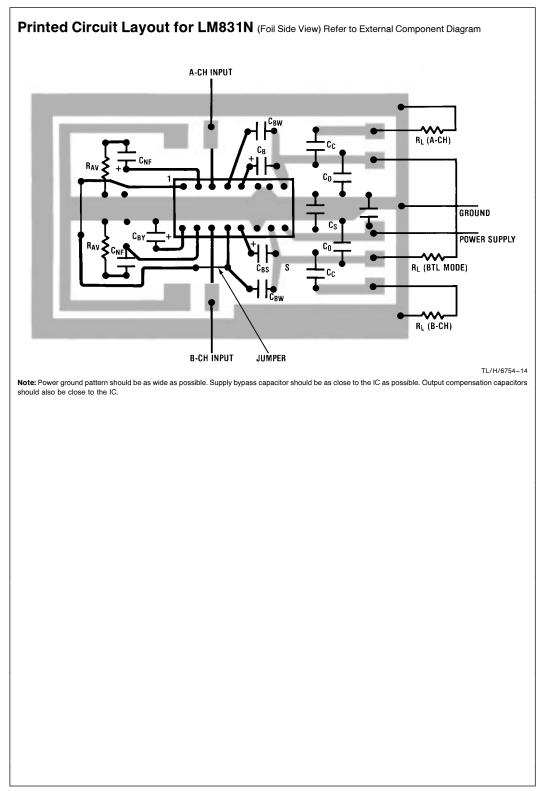
R₅ is provided for bridge (BTL) operation.

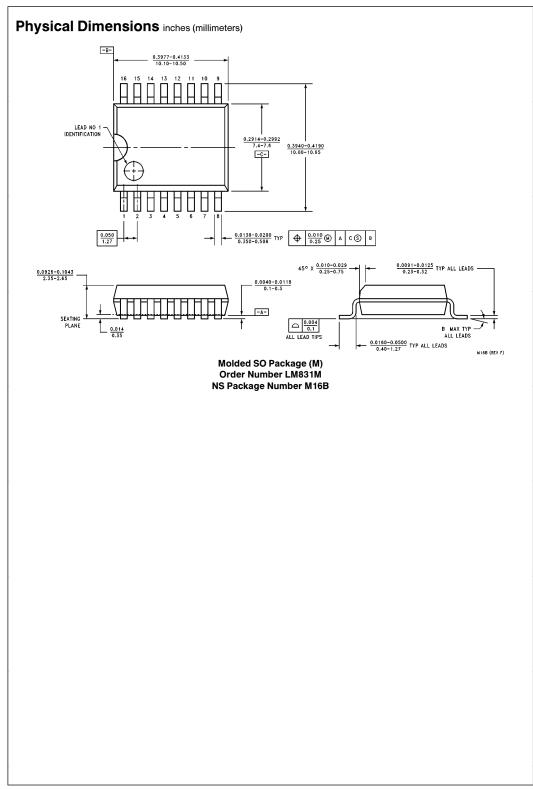
Feedback is provided to the input transistor Q_1 emitter by R_6 and $\mathsf{R}_7.$

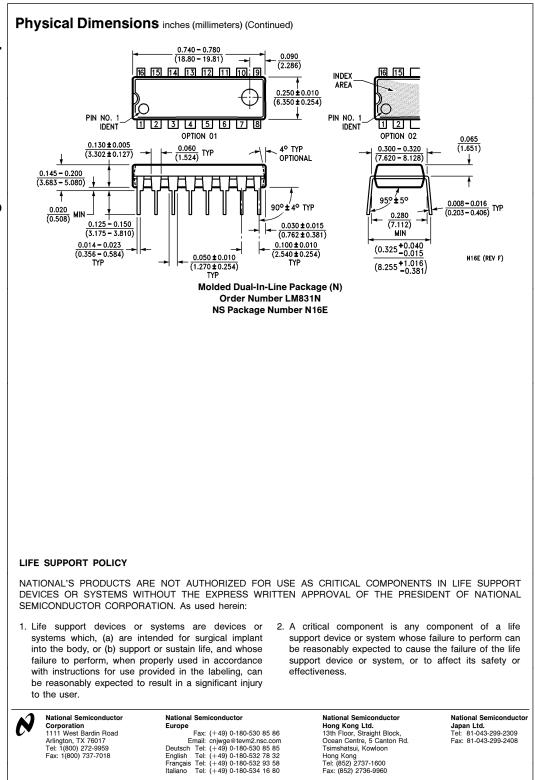




		Comments		Min	Мах	
Co	Required to stabilize output stage.			0.33 μF	1 μF	
C _c	Output coupling the frequency r	g capacitors for Dual Mode. Se esponse. $f_L = \frac{1}{2\pi C_C R_I}$	100 μF	10,000 μF		
C _{BS}	Bootstrap capa Recommended	citors. Sets a low-frequency p	22 μF or (short Pins 4 & 12 to 9)	470 μF		
CS	Supply bypass. reducing supply	Larger values improve low-ba	47 μF	10,000 μF		
C _{BY}		ly for improved low-voltage op	47 μF	470 μF		
C _{NF}	In BTL Mode, C	equency response. Also affects turn-on delay. $f_{L} = \frac{1}{2\pi \cdot C_{NF} \cdot (R_{AV} + 80)}$ $f_{L} = \frac{1}{2\pi \cdot C_{NF} \cdot (R_{AV} + 80)}$ $f_{L} = \frac{1}{2\pi \cdot C_{NF} \cdot (R_{AV} + 80)}$ $f_{L} = \frac{1}{2\pi \cdot C_{NF} \cdot (R_{AV} + 80)}$ $f_{L} = \frac{1}{2\pi \cdot C_{NF} \cdot (R_{AV} + 80)}$				
C _{BTL}	the inverting in	e Bridge Mode. Connects the d out of the other through an inte in one-half the frequency resp $f_L = \frac{1}{2\pi \bullet C_{BTL} \bullet 16}$	ernal resistor. Sets a low- oonse.	0.1 μF 1 μF		
C _{BW}	Improves clipping waveform and sets the high-frequency bandwidth. Works with an internal 16k resistor. (This equation applies for $R_{AV} \neq 0$. For 46 dB application, see BW–C _{BW} curve.) $f_{H} = \frac{1}{2\pi^{\bullet}C_{BW}^{\bullet}16k}$			See table below		
R _{AV}		sed to reduce the gain and improve the distortion and signal to noise. If is is desired, C_{BW} must also be used.			See table below	
			1	C		
Typical A _V		R _{AV}	Min	C _{BW} Max		
46	dB	Short	Open	4700 pF		
40	dB	82	100 pF	4700 pF		
34	dB	240	270 pF	4700 pF		
28 dB		560	500 pF	4700 pF		







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